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# Voltage and Frequency controller for parallel Inverters in Microgrid using Fuzzy logic

# Xuan Hoa Thi Pham<sup>\*</sup>



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#### ABSTRACT

Currently, electrical energy generated from renewable energy sources is being used more and more widely. These energy sources gather and form a microgrid. Microgrid can operate in standalone mode or connected to the grid. When the microgrid operates in standalone mode, it must be controlled to stabilize the frequency and voltage in the microgrid. This article proposes a power control method for inverters in the microgrid, the purpose of the proposed method is to stabilize the frequency and voltage in the microgrid. Besides, the proposed control method also adjusts voltage and frequency to improve power quality in the microgrid. This proposed method is based on fuzzy logic to shift the slope of the droop characteristic curve according to the load. The purpose of the proposed method is to improve the accuracy of power sharing for inverters in microgrids, and it also reduces voltage and frequency deviations in microgrids, it improves power quality in microgrid. The focus of this paper is to improve the voltage quality combined with power sharing between inverters to stabilize the voltage and frequency in microgrid by improving the traditional droop controller, because the droop controller is simple, easy to implement, does not need to use communication network, the disadvantage of the traditional droop controller is the fixed droop coefficient. The proposed controller can adjust the droop coefficient according to the load change based on fuzzy logic inference, the controller is simple, and can be applied to complex microgrids with multiple generators and multiple inverters connected in parallel. The proposed controller is simulated for a islanded microgrid with three parallel connected inverters using Matlab/Simulink software to demonstrate the suitability of the proposed method.

Key words: Droop control, power sharing, voltage control, frequency control, fuzzy logic

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#### History

- Received: 06-02-2023
- Revised: 23-12-2023
- Accepted: 21-9-2024
- Published Online:

#### DOI :

Check for updates

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# **INTRODUCTION**

- <sup>2</sup> Distributed energy sources such as wind energy, so-
- <sup>3</sup> lar energy,... they are sent through power converters.
- <sup>4</sup> There are many different types of microgrid config-
- 5 urations, usually microgrids need an inverter to cre-
- 6 ate AC voltage, the AC voltage at the inverter out-
- 7 put is supplied to AC loads or connected to the grid..
- <sup>8</sup> Power transmission in microgrid achieves high effi<sup>9</sup> ciency when inverters are connected in parallel <sup>1-10</sup>.
- <sup>10</sup> However, in stand-alone mode, the microgrid must
- 11 have power sharing between inverters connected in
- <sup>12</sup> parallel to maintain voltage and frequency stability.
- <sup>13</sup> At the same time, this power sharing will help avoid
- <sup>14</sup> balancing currents flowing between inverters 1-10.
- <sup>15</sup> Based on the power characteristics of the source, pre-
- 16 vious studies have established a mathematical model
- <sup>17</sup> for the droop controller to control the power-sharing
  <sup>18</sup> between inverters operating in parallel. The droop
  <sup>19</sup> controller controls active power according to fre<sup>20</sup> quency and controls reactive power according to volt<sup>21</sup> age. Relationship between active power and fre<sup>22</sup> quency; Reactive power and voltage are expressed

through slope coefficients <sup>1-12</sup>. Therefore, researchers <sup>23</sup> relied on the slope factor to realize power sharing between parallel-connected inverters. However, previous studies often fixed these slope coefficients <sup>1-12</sup>, so <sup>26</sup> when load parameters change a lot, the resulting voltage and frequency at the output of the inverters given <sup>28</sup> by this droop controller will have a much larger deviation from the rated value. <sup>30</sup>

Several studies have presented traditional droop con-31 trol methods for power sharing. The purpose of this study is to share power among inverters without aim-33 ing to reduce voltage and frequency deviation to im-34 prove power quality. However, the traditional droop controller is affected by the line impedance parame-36 ter. Therefore, there have been several studies pre-37 senting improved droop control methods<sup>10-15</sup> for 38 power sharing. However, the purpose of these stud-39 ies is to improve the accuracy of power-sharing for 40 inverters without aiming to reduce voltage and fre-41 quency deviations to improve power quality. 42

Therefore, this paper designs a droop controller com-<br/>bined with fuzzy logic to overcome the disadvantages43of previous controllers.45

**Cite this article :** Pham X H T. **Voltage and Frequency controller for parallel Inverters in Microgrid using Fuzzy logic.** *Sci. Tech. Dev. J. – Engineering and Technology* 2025; ():1-13.

- <sup>46</sup> The contributions of the proposed controller are de-<sup>47</sup> scribed below:
- 48 The droop-fuzzy logic controller automatically ad-
- <sup>49</sup> justs the slope of the droop characteristic curves when
- 50 the load changes. Therefore, this controller will mini-
- <sup>51</sup> mize the frequency and voltage deviation, it improves
- <sup>52</sup> power quality in microgrid.
- 53 In addition, the droop-fuzzy logic controller cor-
- 54 rectly power-sharing between the parallel-connected
- <sup>55</sup> inverters in the Microgrid.
- 56 Typically, the structure of an islanded microgrid con-
- 57 sisting of inverters operating in parallel is shown in
- <sup>58</sup> Figure 1. In standalone mode, the microgrid must be
- <sup>59</sup> capable of self-stabilizing voltage and the frequency
- 60 under load conditions changes in real-time.
- <sup>61</sup> Typically, the traditional power control scheme of an
- <sup>62</sup> independent microgrid includes two control loops:
- The outer control loop is the power control loop.
- 64 Previous studies often used a droop controller
- <sup>65</sup> for this control loop, because the droop con-
- troller is easy to use, does not require com-
- <sup>67</sup> munication, and is easy to change. However,
- 68 the droop controller will give incorrect power
- <sup>69</sup> division results in the case of different line
- <sup>70</sup> impedances, or different inverters. Therefore, in
- this article, we propose a method to improve the
- <sup>72</sup> droop controller by using fuzzy logic.
- The internal control loop is a voltage control
- <sup>74</sup> loop and a current control loop. This control
- <sup>75</sup> loop will control the voltage and current at the
- <sup>76</sup> output of the inverter according to the reference
- voltage and current values.

78 The block diagram of the proposed controller is79 shown in Figure 2.

# 80 METHOD

# 81 Theoretical basis of the proposed control 82 method

<sup>83</sup> The traditional droop controller for dividing power
<sup>84</sup> between inverters is established based on the equiv<sup>85</sup> alent circuit of the inverter connected to the load

<sup>86</sup> through a line as shown in Figure 3.



**Figure 3**: The equivalent circuit of the inverter connects to the load through the line

According to research <sup>1–12</sup>, the power running on the line is calculated according to the expression as Eq. 1 (Figure 23):

90

# In there:

 $\begin{array}{ll} R(\Omega) \mbox{ and } X(\Omega) \mbox{ are impedance parameters of the line.} & {}_{91} \\ V(\mbox{voltage}) \mbox{ is the voltage at the beginning of the line.} & {}_{92} \\ P(W) \mbox{ and } Q(\mbox{Var}) \mbox{ are the powers running on the line.} & {}_{93} \\ V_{PCC}(\mbox{voltage}) \mbox{ is the voltage at the PCC.} & {}_{94} \end{array}$ 

$$\delta$$
 is the phase difference angle of the voltage V anh  $\Psi_{PCC}$ :  $\delta = \delta_1 - \delta_2$ 

$$\dot{Z} = Ze^{j\theta} = R + jX$$

Transforming expression (1), we have: Eq. (2) and Eq. 97 (3) (Figure 23) 96

The actual angle between the output voltage of inverters and PCC bus voltage  $\delta$  is a small value, so  $\sin \delta \approx 100$  $\delta$  and  $\cos \delta = 1$ , X>>R, from equation (2) and (3) we 101 have: Eq. (4) and Eq. (5) (Figure 23) 102

Equations (4) show that: P depends on frequency f. 103 From there, we can establish P/f droop controller to 104 control the active power of the inverters as Eq. 6 (Figure 23): 106

The droop graph of equation (6) is drawn in Figure 4. 107 Figure 4 shows that when the power P changes, the 108 frequency f also changes. If the load increases significantly, the frequency will decrease significantly compared to the rated value and vice versa. 111



Equations (5) show that: Q depends on voltage V. <sup>112</sup> From there, we can establish Q/V droop controller to <sup>113</sup> control the reactive power of the inverters as Eq. (7) <sup>114</sup> (Figure 23): <sup>115</sup>

The droop graph of equation (7) is drawn in Figure 5. 116 Figure 5 shows that when the power Q changes, the 117 voltage V also changes. If the load increases significantly, the voltage will decrease significantly compared to the rated value and vice versa. 120





Figure 2: Block diagram of the proposed controller for an inverter in islanded microgrid

- 121 Where:
- $_{122}$  V<sub>0</sub> is the nominal amplitude voltage.
- $_{123}$  f<sub>0</sub> is the nominal frequency.
- 124 V is the measured amplitude voltage of the inverter.
- 125 f is the measured frequency of the inverter.
- $_{126}$  P<sub>0</sub> is the nominal active power of the inverter.
- $_{\rm 127}~Q_0$  is the nominal reactive power of the inverter.
- $m_p$  and  $m_q$  are the droop coefficients, which are cal-
- 129 culated as Eq. (8) (Figure 23):

# DESIGN OF THE FUZZY LOGIC CONTROLLER

Figure 4 and Figure 5 show that when the power con-132sumption of the load changes, the frequency and volt-133age also change. If the load increases significantly, the134frequency and voltage will decrease significantly com-135pared to the rated value and vice versa. Dividing the136power to the inverters according to the rated ratio is137done by expression (8). We see that expression (8) has138

130

131



a constant value for each inverter, so the slopes of the 139 P/f and Q/V droop curves do not change during op-140 eration. In fact, the load always changes in real time, 141 which will lead to a situation where the voltage and 142 frequency of the microgrid deviate too much from the 143 rated value. This is also something we do not want. 144 Therefore, this article provides a method to change the 145 slope of the drop graph using fuzzy logic. 146

Equations (6), (7) and Figures 4 and 5 show that: a 147 change in the active power of the consumed load ( $\Delta P$ ) 148 will cause a corresponding change in frequency ( $\Delta f$ ). 149 Similarly, a change in the reactive power of the con-150 uming load ( $\Delta Q$ ) will cause a corresponding change 151 in voltage ( $\Delta V$ ). If these changes are significant, it will create unsatisfactory voltage quality in the microgrid. 153 It may also cause balanced currents to flow in the in-154 verters, which will cause damage to the inverters. 155  $V=V_0$  and  $f=f_0$  can only be obtained when  $Q=Q_0$  and 156 157  $P=P_0$ .

The equations (6) and (7) show that the power-sharing 158 for inverters depends on the slope coefficients deter-159 mined in equation (8). Figures 4 and 5 shows that 160 the frequency at the output of the inverter changes ac-161 cording to the active power of the load and the volt-162 age at the output of the inverter changes according to 163 the reactive power of the load; The graph of droop P/f 164 and Q/V in (6) and (7) have slopes that depend on (8). When the slopes (8) change, the power sharing will 166 change accordingly. Therefore, this paper proposes a 167 method to shift the slope coefficients  $m_p$  and  $m_q$  ac-169 cording to changes in the load instead of fixing them according to the equation (8). 170 In the conventional droop method, the slope coeffi-171

<sup>172</sup> cients  $m_p$  and  $m_q$  are fixed according to equation (8). <sup>173</sup> When the load increases or decreases sharply, the fre-<sup>174</sup> quency and voltage at the output of the inverter will deviate much from the value of its norm. In the proposed method, the slope coefficients  $m_p$  and  $m_q$  are adjusted to change according to the load. When the load increases or decreases sharply, the frequency and voltage at the output of the inverter will deviate less than with the conventional droop method.

The slopes  $m_p$  and  $m_q$  of the P/f and Q/V curves are varied by the fuzzy controller. The block diagram for the proposed droop-fuzzy logic controller is shown in Figure 5. Figure 6 is combined from equations (6), (7) and fuzzy-logic block to adjust the slope coefficients  $m_p$  and  $m_q$ . Whereas the conventional droop controller is implemented according to equations (6) and (7),  $m_p$  and  $m_q$  are fixed according to formula (8) as the Figure 7. Therefore, the conventional droop controller will give control results with larger voltage and frequency deviation than the proposed method.

# Design of fuzzy controller to adjust mp 192 slope: 193

<i>This fuzzy controller has two inputs and one output :</i>	194
The first input is the difference between the actual	195
value and the rated value of the active power: $e_p =$	196
$P - P_0$	197
The second input is the rate of change of active power	198
over time: dP/dt	199
The output of this controller is the slip coefficient $m_p$ .	200
Select language variable for input signal:	201
Language variable for $e_p$ input:	202
NB: more negative; NS: less negative; ZE: equal zero;	203
PS: less positive; PB: much positive	204
Language variable for dP/dt input:	205
N: negative; Z: zero; P: positive	206
We can choose more language variables for input and	207
output for better tuning.	208
Select value domain for input signal:	209
Depending on the change of actual load capacity com-	210
pared to the rated capacity value, we choose the ap-	211
propriate value range.	212
Select value domain for $e_p$ input: [-1000 1000]	213
Depending on the rate of increase or decrease of the	214
actual load capacity, we choose the appropriate value	215
range.	216
Select value domain for dP/dt input: [-100 1000]	217
Based on the sliding coefficient calculation expression	218
(8), we choose the value range for the $m_p$ output. Usu-	219
ally the $m_p$ value is very small, so we choose many lan-	220
guage variables for the output to make the adjustment	221
more accurate.	222
Select value domain for moutput: $[0; 5.e^{-4}]$	223
Language variable for m output:	224
A1, A2, A3: small; B1, B2, B3: medium; C1, C2, C3:	225
big	226





227 Choose membership functions for the inputs:

We can choose a membership function of triangular
or trapezoidal shape for the input as shown in Figures 8 and 9. We can choose a bar-shaped membership function for the output as shown in Figure 10.



N Z P -100 -50 50 100

Figure 9: Membership function of input dP/dt



- 232 Choose membership functions for the output :
- 233 We can choose a membership function that is trian-
- 234 gular, trapezoidal, or linear.

# 235 Design of fuzzy controller to adjust m

<sup>236</sup> *This fuzzy controller has two inputs and one output :* 

The first input is the difference between the actual  $_{237}$  value and the rated value of the reactive power:  $e_q = _{238}$  Q - Q<sub>0</sub>  $_{239}$ 

- 240 The second input is the rate of change of reactive
- 241 power over time: dQ/dt
- <sup>242</sup> The output of this controller is the slip coefficient  $m_q$ .
- 243 Select language variable for input signal:
- <sup>244</sup> Language variable for  $e_q$  input:
- 245 NB: more negative; NS: less negative; ZE: equal zero;
- <sup>246</sup> PS: less positive; PB: much positive
- 247 Language variable for dQ/dt input:
- 248 N: negative; Z: zero; P: positive
- 249 We can choose more language variables for input and
- <sup>250</sup> output for better tuning.
- 251 Select value domain for input signal:
- 252 Depending on the change of actual load capacity com-
- 253 pared to the rated capacity value, we choose the ap-
- 254 propriate value range.
- 255 Select value domain for  $e_q$  input: [-1000 1000]
- 256 Depending on the rate of increase or decrease of the
- 257 actual load capacity, we choose the appropriate value
- 258 range.
- 259 Select value domain for dQ/dt input: [-1000 1000]
- 260 Depending on the rate of increase or decrease of the
- 261 actual load capacity, we choose the appropriate value
- 262 range.
- 263 Select value domain for dQ/dt input: [-100 1000]
- 264 Based on the sliding coefficient calculation expression
- $_{265}$  (8), we choose the value range for the m<sub>q</sub> output. Usu-
- $_{266}$  ally the m<sub>p</sub> value is very small, so we choose many lan-
- 267 guage variables for the output to make the adjustment
- 268 more accurate.
- <sup>269</sup> Select value domain for moutput:  $[0; 5.e^{-4}]$
- 270 Language variable for m output:
- <sup>271</sup> A1, A2, A3: small; B1, B2, B3: medium; C1, C2, C3: <sup>272</sup> big
- 273 Choose membership functions for the inputs:
- <sup>274</sup> We can choose a membership function of triangular <sup>275</sup> or trapezoidal shape for the input as shown in Fig-<sup>276</sup> ure 11 and Figure 12. We can choose a bar-shaped <sup>277</sup> membership function for the output as shown in Fig-<sup>278</sup> ure 13.
  - NB NS ZE PS PB\_\_\_\_\_

**Figure 11**: Membership function of input  $e_q$ 

279 Choose membership functions for the output :

280 We can choose a membership function that is trian-281 gular, trapezoidal, or linear.



Figure 12: Membership function of input dQ/dt





282

#### **Define control rules:**

Equations (6), (7), (8) and Figures 4 and 5 show them: 283 If  $Q > Q_0$  then  $V < V_0$ . In order for V to get close to 284  $V_0$ , we have to control to decrease the slope mq, which 285 means we have to control to increase the angle  $\alpha$ . 286 If  $Q < Q_0$  then  $V > V_0$ . In order for V to get close to 287  $V_0$ , we have to control to decrease the slope mq, which 288 means we have to control to increase the angle  $\alpha$ . If  $P > P_0$  then  $f < f_0$ . In order for f to get close to  $f_0$ , we 290 have to control to decrease the slope mp, which means 291 we have to control to increase the angle  $\alpha$ . 202 If  $P < P_0$  then  $f > f_0$ . In order for f to get close to  $f_0$ , we 293 have to control to decrease the slope mp, which means 294 we have to control to increase the angle  $\alpha$ . 295 On the other hand, we rely on the language variables, 296

the range of values, the membership function of the 297 input and the output, we can set up the control rules 298 as shown in Table 1. 299

If  $e_p = NB (P \ll P_0)$  and = N (P is decreasing) then we so choose  $m_p$  output as A1.

If  $e_p = NS (P \le P_0)$  and = N (P is decreasing) then we 302 choose  $m_p$  output as B1. 303

If  $e_p = ZE (P = P_0)$  and = N (P is decreasing) then we 304 choose  $m_p$  output as C1. 305

If  $e_p = PS$  and dP/dt = N then we choose  $m_p$  output 306 as B3. 307

If  $e_p = BP$  and dP/dt = N then we choose  $m_p$  output 308 as A3. 309

If  $e_p = NB$  and dP/dt = Z then we choose  $m_p$  output <sup>310</sup> as A2. <sup>311</sup> <sup>312</sup> If  $e_p = NB$  and dP/dt = P then we choose  $m_p$  output <sup>313</sup> as A3.

<sup>314</sup> If  $e_p = NS$  and dP/dt = Z then we choose  $m_p$  output <sup>315</sup> as B2.

- <sup>316</sup> If  $e_p = ZE$  and dP/dt = Z then we choose  $m_p$  output <sup>317</sup> as C2.
- <sup>318</sup> If  $e_p = PS$  and dP/dt = Z then we choose  $m_p$  output as <sup>319</sup> B2.
- <sup>320</sup> If  $e_p = PB$  and dP/dt = Z then we choose  $m_p$  output <sup>321</sup> as A2.
- <sup>322</sup> If  $e_p = NS$  and dP/dt = P then we choose  $m_p$  output <sup>323</sup> as B3.
- <sup>324</sup> If  $e_p = ZE$  and dP/dt = P then we choose  $m_p$  output <sup>325</sup> as C3.
- <sup>326</sup> If  $e_p = PS$  and dP/dt = P then we choose  $m_p$  output as <sup>327</sup> B1.
- <sup>328</sup> If  $e_p = PB$  and dP/dt = P then we choose  $m_p$  output as <sup>329</sup> A1.
- $_{330}$  Similarly, we can choose the control law for the fuzzy  $_{331}$  controller m<sub>q</sub>

332 Choose the composition rule according to the Sum-

<sup>333</sup> Prod principle. Defuzzification by the centroid<sup>334</sup> method.

#### 335 Calculation and low-pass filter

336 The instantaneous power p and q are calculated ac-

337 cording to the instantaneous value of the current flow-

 $_{\rm 338}\,$  ing on the line (i\_2) and the voltage at the beginning of

 $_{339}$  the line (v<sub>c</sub>) in a stationary *dq0* frame as Eq. (9) and  $_{340}$  (10) (Figure 23):

### <sup>341</sup> The current controller and voltage con-<sup>342</sup> troller

343 Based on the equivalent circuit shown in Figure 14,

<sup>344</sup> we set up these controllers.

345 Where:

 $_{346}$  R ( $\Omega$ ) and L (H) are resistor and inductance of the line.

 $_{^{347}}$  R<sub>f</sub> ( $\Omega$ ) and L<sub>f</sub> (H) are resistor and inductance of the  $_{^{348}}$  filter (H).

<sup>349</sup> Applying Kirchhoff's laws to the circuit of Figure 14, <sup>350</sup> we have the following equations (11) and (12) (Fig-<sup>351</sup> ure 23):

- <sup>352</sup> Projecting equation (12) onto the coordinate system <sup>353</sup> dq0. The voltage bias is eliminated by the PI con-<sup>354</sup> troller on both the d-axis and the q-axis, the d-axis <sup>355</sup> voltage bias is added with a feedback component (-<sup>356</sup>  $\omega Cv_{ca}$ ), and a voltage bias on the q-axis is added feed-
- <sup>357</sup> back component( $\omega Cv_{cd}$ ). According to [15], the ca-

<sup>358</sup> pacitance of the filter capacitor is usually very small, <sup>359</sup> so the C  $dv_{cd}/dt$  and C  $dv_{cq}/dt$  components are ig-<sup>360</sup> nored, the equations (11) can be written as Equation <sup>361</sup> (13) and (14) (Figure 23). Equations (13) and (14) are the voltage controller. 362 Projecting equation (12) onto the coordinate system 363 dq0. The current bias is eliminated by the PI controller on both the d-axis and the q-axis, the d-axis 365 current bias is added with a feedback component (- $\omega L_f i_{1q}$ ), and a current bias on the q-axis is added 367 feedback component ( $\omega L_f i_{1d}$ ). According to [15], 368 the inductance and resistance of the filter capacitor 369 is usually very small, so the ( $L_f$  (di<sub>1d</sub>)/dt +  $R_f$  i<sub>1d</sub>) 370 and ( $L_f$ (di<sub>1q</sub>)/dt +  $R_f i_{1q}$ ) components are ignored, 371 the equations (12) can be written as equation (15) and 372 (16) (Figure 23): 373

Equations (15) and (16) are the current controller.

374

375

376

383

# SIMULATION RESULT AND DISCUSSION

Simulate power control for a microgrid consisting 377 of three inverters connected in parallel using Matlab/simulink software. Simulations are performed by 379 the conventional controller and the proposed controller to compare the results. Simulation parameters 381 are given in Table 1. 382

#### Case 1:

Power sharing simulation for two inverters in an independent microgrid is performed using the proposed controller. 386





Figure 15a and 15b show that for an erroneous value  $_{387}$  of active power (e<sub>p</sub>) at the input of the fuzzy logic  $_{388}$ 



Parameters	Values	Parameters	Values
DC link voltage Vcd (V)	600	Nominal frequency $f_0$ (Hz)	50
Filter $L_f$ (mH)	4.2	Nominal power (kVA)	4
Filter $\mathbf{R}_f$ ( $\mathbf{\Omega}$ )	0.1	Nominal voltage $V_{AC,p}$ (V)	310
Filter C (mF)	2.2	m <sub>q</sub> (V/Var)	1.05e-4
$f_z(kHz)$	10	$m_p (rad/s /W)$	1e-4
k <sub>pv</sub>	0.1	k <sub>pi</sub>	10
K <sub>iv</sub>	0.05	k <sub>ii</sub>	1
Line impedances			
L <sub>1</sub> (mH)	1	$R_1(\Omega)$	1.2
L <sub>2</sub> (mH)	0.8	$R_2(\Omega)$	0.9
L <sub>3</sub> (mH)	0.5	$R_3(\Omega)$	0.7

#### **Table 1: SIMULATION PARAMETERS**



<sup>389</sup> block, there will be a corresponding slope coefficient <sup>390</sup>  $m_p$  at the output of the fuzzy logic block. The value <sup>391</sup> of  $m_p$  is within the specified range selected (the range <sup>392</sup> of values is selected by formulas (6), (7), and (8)). On <sup>393</sup> the other hand, when the  $e_p$  deviation decreases,  $m_p$ <sup>394</sup> increases (increases in the selected value range), and <sup>395</sup> vice versa when the  $e_p$  deviation increases (from 6s <sup>396</sup> onwards) the  $m_p$  decreases. These results are completely consistent with the fuzzy control law estab- 397 lished. Figure 15c shows that when the slope coeffi- 398 cient  $m_p$  changes, the frequency at the output of the 399 Droop-fuzzy logic controller or the output frequency 400 of the inverter also changes accordingly, the curve of 401 frequency shows as the load increases then the fre- 402 quency decreases, which is also completely consistent 403 with the equation (6) and power curve in the Figure 4. 404 The f value is also calculated according to  $m_p$  and  $e_p$  in 405 equations (7), (8), and the frequency deviation from 406 the norm is calculated in Table 2. Table 2 shows that 407 the frequency deviation is very small, this result is due  $_{\rm 408}$ to the fuzzy-logic controller that controls to change 409  $m_p$  when the load changes. As we see in the Figure 4, 410 if we don't change the slope of the droop curve P/f as 411 the load changes, then a power deviation  $\Delta P$  will give a 412 corresponding frequency deviation  $\Delta f$ , and when the 413 load changes more, then the  $\Delta f$  will be large. Same 414 as above, Figures 15d and 15e show that for a devia- 415 tion of reactive power  $e_q$ , it will give a value of slope 416



Figure 15: The graph showing the change of slip coefficients (mp, mq) and frequency and voltage as the load increases

coefficient  $m_q$  in the selected range of values, and we 417 see that when the deviation  $e_q$  decreases, the slope co-418 efficient  $m_q$  increases and vice versa, at t=6s, the  $e_q$ 419 deviation continues to decrease compared to the pre-420 vious one, so the slope coefficient  $m_a$  at continues to 421 increase. This is completely consistent with the fuzzy 422 control law established. Figure 15f shows that when 423 the slope coefficient  $m_q$  changes, the voltage at the 424 output of the Droop-fuzzy logic controller or the out-425 put voltage of the inverter also changes accordingly, 426 427 which is also completely consistent with equation (7) and power curve in Figure 4b. The voltage value is 428 also calculated according to  $m_q$  and  $e_q$  in formula (7), (8), voltage deviation from the norm is calculated in 430 Table 2. Table 2 shows that the voltage difference is 431 very small, to get this result is because we control the 432 433 slope of the power characteristic curve when the load changes. As we see in Figure 5, if we do not change the 434 slope of the power curve Q/V when the load changes, 435 then a power deviation  $\Delta Q$  will give a correspond-436 ing voltage deviation  $\Delta V$ , and when the load changes 437  $_{438}$  more, then the  $\Delta V$  will be large.

<sup>439</sup> Figures 16 and 17 show the Droop-fuzzy logic con-<sup>440</sup> trol method for a good dynamic response as soon as the load changes and the current and voltage stabilizes 441 well right after the load changes. The output power of 442 the inverters is divided exactly in a 1:1 ratio. 443 The simulation results above show that the proposed 444 method controls the correct sharing power for the in- 445 verters and reduces the voltage and frequency devia- 446 tion when the load changes suddenly, improving the 447 quality of the power supply for the load, and a good 448 dynamic response. The proposed method has over- 449 come the disadvantages of the conventional or im- 450 proved droop methods because the slope of this power 451 curve is always fixed, so when the load changes, the 452 voltage and frequency deviations cannot be adjusted 453 to improve power quality. 454

#### Case 2:

Two inverters have the same rated power, they are 456 connected in parallel, Simulate power division by the 457 proposed method. 458

Figure 18 shows that the proposed controller gives459good power division results, the power division re-<br/>sults in the time period from 0s to 5s have the follow-<br/>ing deviations:460

455

able 2: Table of results for frequency and voltage deviation from rated value					
Load parameters	$(\Delta f=f0-f)$	$(\Delta V=V0-V)$			
Z1 = 20+j3 (W)	$\Delta f$ =50-50.0276 = -0.276(Hz)	ΔV=310-310.0016 = -0.0016(V)			
Z2 = 10+j2 (W)	$\Delta f$ =50-49.9686 = 0.0314(Hz)	$\Delta V = 310 - 309.98 =$ 0.002 (V)			



Figure 18: The graph of sharing active power, reactive power, and voltage at load when simulated by the proposed droop-fuzzy controller

 $_{463} e_p \% = (P_i - P^*_i)/(P^*_i).100\% = (1118 _{464} 1110)/(1110).100\% = 0.72 \%$ 

<sup>465</sup> The error of sharing the reactive power of inverters for<sup>466</sup> the period from 0s to 5s:

 $e_q \% = (Q_i - P_i^*)/(Q_i^*).100\% = (625-620)/(620).100\%$   $e_{68} = 0.8 \%$ 

<sup>469</sup> This shows that the proposed method gives highly ac-<sup>470</sup> curate power division results.

471 Figure 19 shows that the conventional droop con-

472 troller gives poor power-sharing results, the error is

473 very large when sharing reactive power. The error of

<sup>474</sup> sharing the reactive power of inverters for the period <sup>475</sup> from 0s to 5s:

<sup>476</sup>  $e_q \% = (Q_i - P_i^*)/(Q_i^*).100\% = (644-620)/(620).100\%$ <sup>477</sup> = 3.8 %

478 The voltage graph in Figure 18 also shows that the
479 voltage deviation given by the proposed drooping
480 fuzzy controller is very small, especially as the load
481 increases, the voltage drop is smaller than the voltage
482 graph in Figure 19.

#### Case 3:

# 483

492

500

Use the proposed method to divide the power between the two inverters according to the rated power 485 ratio of 2:1,  $P_{dm1}=2P_{dm2}$  486 Figure 20 shows that the proposed controller has 487 given the correct power-sharing results with the rated 488 power ratio 2:1, its transient response and steady-state 489

response are also very good, and the time steady-state 490 set up early. 491

#### Case 4:

Use the proposed method to divide the power between the three inverters according to the rated power 494 ratio of 1:1:1,  $P_{dm1}=P_{dm2}=P_{dm3}$ . 495 Figure 21 shows that the proposed controller gave the 496 correct power sharing results with the rated power ra-

correct power-sharing results with the rated power ra-497tio 1:1:1, its steady-state response is very good, and the498time steady-state is set up early.499

# Case 5:

Use the proposed method to divide the power among 501 the three inverters according to the rated power ratio 502















Figure 21: The graph of sharing active power, reactive power, and voltage at load by the proposed droop-fuzzy controller

<sup>503</sup> of 1:1:1, the load varying in this case.

504 Figure 22 shows that the proposed controller gave the

505 correct power-sharing results with the rated power ra-

<sup>506</sup> tio even when the load has changed.

507 The above simulation results show that the proposed

<sup>508</sup> controller gives good results, negligible deviation, and<sup>509</sup> relatively early setup time. Good voltage quality, low

510 voltage and frequency deviation.

# **S11 CONCLUSION**

<sup>512</sup> In this paper, a sliding coefficient adjustment method using fuzzy logic is proposed, improving the tradi-513 tional droop controller. This innovation allows to sig-514 nificantly improve the accuracy of power sharing for 515 inverters in the microgrid, improve the quality of volt-516 age supplied to loads, and reduce frequency deviation. 517 The advantages of the proposed method were verified 518 by simulation for microgrids with two and three in-519 520 verters.

# 521 CONFLICT OF INTEREST

522 There is no conflict of interest regarding this 523 manuscript.

#### AUTHOR CONTRIBUTION

Xuan Hoa Thi Pham have writen this entire paper.

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$\widetilde{S} = \dot{V}.I^* = \dot{V}.\left(\frac{\dot{V} \cdot \dot{V}_{PCC}}{\dot{Z}}\right)^* = V.e^{j\delta_1}\left(\frac{Ve^j}{d}\right)^*$	$\frac{i\delta_1 - V_{PCC} e^{j\delta_2}}{Z e^{j\theta}} \bigg)^* = P + jQ  (1)$	$p = \frac{3}{2} (i_{2d} v_{cd} + i_{2q} v_{cq})$	(9)
$\sin \delta = \frac{\text{XP-RQ}}{\text{VV}_{\text{PCC}}}$ $\text{V-V}_{\text{PCC}}\cos \delta = \frac{\text{RP+XQ}}{\text{C}}$	(2)	$q = \frac{1}{2} \left( i_{2d} \mathbf{v}_{cq} \cdot \mathbf{i}_{2q} \mathbf{v}_{cd} \right)$ $\begin{cases} i_1 = i_2 + C \frac{d \mathbf{v}_c}{dt} + \mathbf{i}' \\ d\mathbf{i} \\ \end{array}$	(10) (11)
$\delta = \frac{XP}{VV_{PCC}}$ $V_{} = \frac{XQ}{V}$	(4)	$ \begin{pmatrix} v_{inv} = L_f \frac{\omega_1}{dt} + R_f i_1 + v_c \\ i_{1d} = \\ k_{ov} (v_{od}^* - v_{od}) + k_{iv} \int (v_{od}^* - v_{od}) dt + i_{2d} - \omega C v_{cq} \end{pmatrix} $	(12)
$f = f_0 - m_p(P - P_0)$ $V = V_{-} m_p(Q - Q_0)$	(6)	$\begin{cases} i_{1q} = \\ k_{pv}(v_{cq}^* - v_{cq}) + k_{iv} \int (v_{cq}^* - v_{cq}) dt + i_{2q} + \omega C v_{cd} \end{cases}$	(14)
$m_{p} = \frac{f_{0} - f_{min}}{P_{max} - P_{0}};  m_{q} = \frac{V_{0} - V_{min}}{Q_{max} - Q_{0}}$	(8)	$\begin{cases} v_{invd}^{v_{invd}^{-}} \\ k_{pi}(i_{1d}^{*} - i_{1d}) + k_{ii} \int (i_{1d}^{*} - i_{1d}) dt & -\omega L_{f} i_{1q} + v_{cd} \\ v_{invq}^{-} \\ \end{cases}$	(15)
		$\Big( k_{pi} (i_{1q}^* \cdot i_{1q}) + k_{ii} \int (i_{1q}^* \cdot i_{1q}) dt + \omega L_f i_{1d} + v_{cq}$	(16)

Figure 23: Equation

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# Bộ điều khiển điện áp và tần số cho các bộ nghịch lưu kết nối song song trong microgrid sử dụng logic mờ

# Phạm Thị Xuân Hoa<sup>\*</sup>



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# TÓM TẮT

Hiên nay, năng lương điên được tao ra từ các nguồn năng lượng tái tao đang được sử dụng ngày càng rộng rãi. Các nguồn năng lượng này tập hợp lại và hình thành nên một lưới điện siêu nhỏ. Lưới điện siêu nhỏ có thể hoạt động ở chế độ độc lập hoặc được kết nối với lưới điện. Khi lưới điện siêu nhỏ hoạt động ở chế độ độc lập, nó phải được điều khiển để ổn định tần số và điện áp trong lưới điện siêu nhỏ. Bài viết này đề xuất một phương pháp điều khiển công suất cho các bộ nghịch lưu trong lưới điện siêu nhỏ, mục đích của phương pháp đề xuất là ổn định tần số và điện áp trong lưới điện siêu nhỏ. Bên cạnh đó, phương pháp điều khiển đề xuất cũng điều chỉnh điện áp và tần số để cải thiện chất lượng điện năng trong lưới điện siêu nhỏ. Phương pháp đề xuất này dựa trên logic mờ để dịch chuyển độ dốc của đường cong đặc tính droop theo tải. Mục đích của phương pháp đề xuất là cải thiện độ chính xác của việc chia sẻ công suất cho các bộ nghịch lưu trong lưới điện siêu nhỏ và nó cũng làm giảm độ lệch điện áp và tần số trong lưới điện siêu nhỏ, nó cải thiện chất lượng điện năng trong lưới điện siêu nhỏ. Trọng tâm của bài báo này là cải thiện chất lượng điện áp kết hợp với chia sẻ công suất giữa các bộ nghịch lưu để ổn định điện áp và tần số trong lưới điện siêu nhỏ bằng cách cải tiến bộ điều khiển droop truyền thống, vì bộ điều khiển droop đơn giản, dễ triển khai, không cần sử dụng mạng truyền thông, nhược điểm của bộ điều khiển droop truyền thống là hê số droop cố đinh. Bô điều khiển được đề xuất có thể điều chỉnh hê số droop theo sư thay đổi tải dưa trên suy luân logic mờ, bô điều khiển đơn giản và có thể áp dụng cho các lưới điện siêu nhỏ phức tạp với nhiều nguồn điện và nhiều bộ nghịch lưu được kết nối song song. Bộ điều khiển đề xuất được mô phỏng cho một lưới điện siêu nhỏ độc lập với ba bộ biến tần được kết nối song song bằng phần mềm Matlab/Simulink để chứng minh tính phù hợp của phương pháp được đề xuất.

Từ khoá: Điều khiển droop, chia sẻ công suất, điều khiển điện áp, điều khiển tần số, logic mờ

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#### Lịch sử

- Ngày nhận: 06-02-2023
- Ngày sửa đổi: 23-12-2023
- Ngày chấp nhận: 21-9-2024
- Ngày đăng:

#### DOI :



### Bản quyền

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Trích dẫn bài báo này: Hoa P T X. Bộ điều khiển điện áp và tần số cho các bộ nghịch lưu kết nối song song trong microgrid sử dụng logic mờ. Sci. Tech. Dev. J. - Eng. Tech. 2025; ():1-1.